19-4806; Rev 1; 4/10 EVALUATION KIT AVAILABLE



General Description

Features

The MAX9830 mono 2W Class D amplifier provides Class AB audio performance with Class D efficiency.

Active emissions limiting edge rate and overshoot control circuitry greatly reduces EMI. A patented filterless spread-spectrum modulation scheme eliminates the need for output filtering found in traditional Class D devices. These features reduce application component count.

The MAX9830's industry-leading 1.6mA at 5V, 1.2mA at 3.6V, guiescent current extends battery life in portable applications.

The MAX9830 is available in an 8-pin TDFN (2mm x 2mm x 0.8mm) and is specified over the extended -40°C to +85°C temperature range.

Applications

Notebook and Netbook Computers Cellular Phones MP3 Players Portable Audio Players **VoIP** Phones

8 :

1

IN+

PVDD OUT+ OUT- PGND 7 6 5

///XI///

MAX9830

2 3 4

IN-

*EP = EXPOSED PAD. CONNECT THE EP TO PGND TO ENHANCE THERMAL DISSIPATION.

TDFN 2mm x 2mm x 0.8mm

*EP

SHDN N.C.

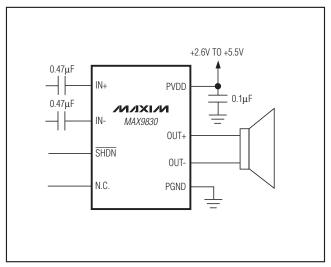
- Industry-Leading Quiescent Current: 1.6mA at 5V, 1.2mA at 3.6V
- Spread Spectrum and Active Emissions Limiting
- Passes EMI Limit Unfiltered with Up to 24in (61cm) of Speaker Cable
- Click-and-Pop Suppression
- Thermal and Overcurrent Protection
- Low 0.5µA Current Shutdown Mode
- Space-Saving, 2mm x 2mm x 0.8mm, 8-Pin TDFN Package

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE		
MAX9830AETA+	-40°C to +85°C	8 TDFN-EP*		

+Denotes a lead(Pb)-free/RoHS-compliant package. *EP = Exposed pad.

Typical Operating Circuit



Pin Configuration

M/X/M

TOP VIEW

Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

ABSOLUTE MAXIMUM RATINGS

Voltage

PVDD, IN+, IN-, SHDN, to PGND	-0.3V to +6V
OUT+, OUT- to PGND0.3V to V	
Current	
Continuous Current Into/Out of PVDD, PGND,	
OUT+, OUT	±600mA
Continuous Input Current (all other pins)	±20mA
Duration of Short Circuit Between	
OUT+, OUT-, and PVDD, PGND	Continuous

Continuous Power Dissipation for a Multilaye $(T_A = +70^{\circ}C)$	r Board
8-Pin TDFN-EP (derate 11.9mW/°C)	953.5mW
Junction Temperature	+150°C
Operating Temperature Range	40°C to +85°C
Storage Temperature Range	65°C to +150°C
Lead Temperature (soldering, 10s)	
Soldering Temperature (reflow)	+260°C
Rate of Voltage Rise at PVDD	1V/µs

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

ELECTRICAL CHARACTERISTICS

 $(V_{PVDD} = V_{SHDN} = 5V, V_{PGND} = 0V, R_L = \infty$, unless otherwise specified. R_L connected between OUT+ and OUT-, AC measurement bandwidth 20Hz to 22kHz, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at $T_A = +25^{\circ}$ C.) (Notes 1, 2)

PARAMETER	SYMBOL	CON	MIN	ТҮР	MAX	UNITS	
SPEAKER AMPLIFIER	•			•			•
Voltage Range	PVDD	Inferred from PSRR	2.6		5.5	V	
		V _{PVDD} = 5.0V			1.6	2.5	mA
Quiescent Supply Current	IDD	V _{PVDD} = 3.6V			1.2		
Shutdown Supply Current	ISHDN	$V_{\overline{SHDN}} = 0V, T_A = +$	25°C		0.5	10	μA
Turn-On Time	ton				1.9	4	ms
Bias Voltage	VBIAS				1.31		V
	Max	Single ended			1		
Maximum AC Input Voltage	VIN	Differential			2		VRMS
Input Decistores in Chutdown	Duise	Between inputs		85.6		kΩ	
Input Resistance in Shutdown	RINSD	From inputs to PGNI		43			
Input Resistance	RIN		12	20		kΩ	
Voltage Gain	Av				12		dB
Common-Mode Rejection Ratio	CMRR	f _{IN} = 1kHz, input refe		48		dB	
	PSRR	$V_{PVDD} = 2.6V \text{ to } 5.5$	54	64.3		dB	
Power-Supply Rejection Ratio		PVDDRIPPLE =	$f_{RIPPLE} = 217Hz$		72		
		200mV _{P-P} (Note 3)	$f_{RIPPLE} = 20 kHz$	64			
Output Power	Dour	THD+N = 10%,	$R_L = 8\Omega$	1.5		w	
	Pout	$f_{IN} = 1 \text{kHz}$	$R_L = 4\Omega$		2.25		vv
Total Harmonic Distortion Plus		f _{IN} = 1kHz	$R_L = 8\Omega, P_{OUT} = 0.5W$		0.04		%
Noise			$R_L = 4\Omega$, $P_{OUT} = 1W$	0.04		1	
Output Offset Voltage	VOS	$T_A = +25^{\circ}C$			±3	±30	mV
Click and Pop Loval	d Dara Laural	Peak voltage, A-weighted, 32	Into shutdown		-56		dBV
Click-and-Pop Level	K _{CP}	samples/sec (Notes 3, 4)	Out of shutdown		-56		UDV

ELECTRICAL CHARACTERISTICS (continued)

 $(V_{PVDD} = V_{SHDN} = 5V, V_{PGND} = 0V, R_L = \infty$, unless otherwise specified. R_L connected between OUT+ and OUT-, AC measurement bandwidth 20Hz to 22kHz, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.) (Notes 1, 2)

PARAMETER	SYMBOL	CONDITIONS	MIN TYP	МАХ	UNITS
Oscillator Frequency	fosc		600		kHz
Spread-Spectrum Bandwidth			±10		kHz
Noise	V _N	A-weighted (Note 3)	39		μV _{RMS}
Signal-to-Noise Batio SNB POUT =		P_{OUT} = P_{OUT} at 1% THD+N, A-weighted R_L = 8Ω	98		dB
Output Current Limit	ILIM	$T_A = +25^{\circ}C$	3		Α
Thermal Shutdown Level			+180)	°C
Thermal Shutdown Hysterysis			30		°C
Efficiency	η	$R_L = 8\Omega$, $P_{OUT} = 1.5W$	85		%
DIGITAL INPUT (SHDN)					
Input Voltage High	VIH		1.4		V
Input Voltage Low	VIL			0.4	V
Input Leakage Current		$T_A = +25^{\circ}C$		±10	μA

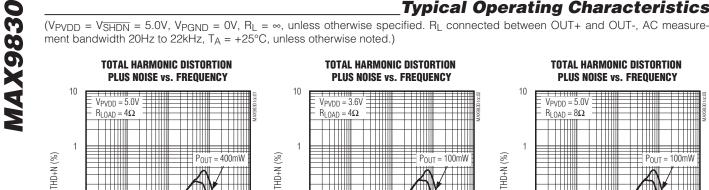
Note 1: All devices are 100% production tested at $T_A = +25^{\circ}C$. All temperature limits are guaranteed by design.

Note 2: Testing performed with a resistive load in series with an inductor to simulate an actual speaker load. For $R_L = 4\Omega$, $L = 33\mu$ H. For $R_L = 8\Omega$, $L = 68\mu$ H.

Note 3: Amplifier inputs AC-coupled to PGND with $C_{IN} = 0.47 \mu F$.

Note 4: Specified at room temperature with an 8Ω resistive load in series with a 68μ H inductive load connected across BTL outputs. Mode transitions are controlled by SHDN.

TOTAL HARMONIC DISTORTION



PLUS NOISE vs. FREQUENCY PLUS NOISE vs. FREQUENCY PLUS NOISE vs. FREQUENCY 10 V_{PVDD} = 5.0V 10 V_{PVDD} = 3.6V $V_{PVDD} = 5.0V$ $R_{IOAD} = 4\Omega$ $R_{IOAD} = 8\Omega$ 1 1 THD+N (%) POUT = 400mW (%) N+DH E P_{OUT} = 100mW POUT = 100mW 0.1 0.1 0.1 POUT = 600mW $P_{OUT} = 1.4W$ 1W POUT 1 1 1 11 111 0.01 0.01 0.01 10 100 1k 100k 100 1k 100k 10 100 1k 10k 10 10k 10k FREQUENCY (Hz) FREQUENCY (Hz) FREQUENCY (Hz) **TOTAL HARMONIC DISTORTION TOTAL HARMONIC DISTORTION TOTAL HARMONIC DISTORTION** PLUS NOISE vs. OUTPUT POWER **PLUS NOISE vs. OUTPUT POWER PLUS NOISE vs. FREQUENCY** 10 10 10 V_{PVDD} = 3.6V V_{PVDD} = 3.6V $V_{PVDD} = 5.0V$ $R_{IOAD} = 8\Omega$ $R_{IOAD} = 4\Omega$ $R_{IOAD} = 4\Omega$ 1 1 1 THD+N (%) f = 1kHz = f = 1kHz THD+N (%) $P_{OUT} = 100 \text{mW}$ THD+N (%) = 6kHz f = 6kHz0.1 0.1 0.1 $P_{OUT} = 500 \text{mW}$ = 100Hz = 100Hz 0.01 0.01 0.01 10 100 1k 10k 100k 0 0.5 1.0 1.5 20 25 3.0 0 0.2 0.4 0.6 0.8 1.0 1.2 OUTPUT POWER (W) FREQUENCY (Hz) OUTPUT POWER (W) TOTAL HARMONIC DISTORTION TOTAL HARMONIC DISTORTION PLUS NOISE vs. OUTPUT POWER **PLUS NOISE vs. OUTPUT POWER OUTPUT POWER vs. LOAD RESISTANCE** 10 10 2.5 $V_{PVDD} = 5.0V$ $V_{PVDD} = 3.6V$ $V_{PVDD} = 5.0V$ $\mathsf{R}_{LOAD}=8\Omega$ $R_{LOAD}=8\Omega$ 2.0 **OUTPUT POWER (W)** 1 1 THD+N (%) THD+N (%) 1.5 THD+N = 10% f = 6 kHzf = 6kHz 1.0 = 1kH7 1kHz 0.1 0.1 0.5 THD+N = 1%100H; 100Hz 0.01 0.01 0 0 0.2 0.4 0.6 0.8 1.0 1.2 1.4 1.6 1.8 0.1 0.2 0.3 0.4 0.5 0.6 0.7 0.8 0.9 0 1 10 OUTPUT POWER (W) OUTPUT POWER (W) LOAD RESISTANCE (Ω)

TOTAL HARMONIC DISTORTION

/N/IXI/N

Typical Operating Characteristics

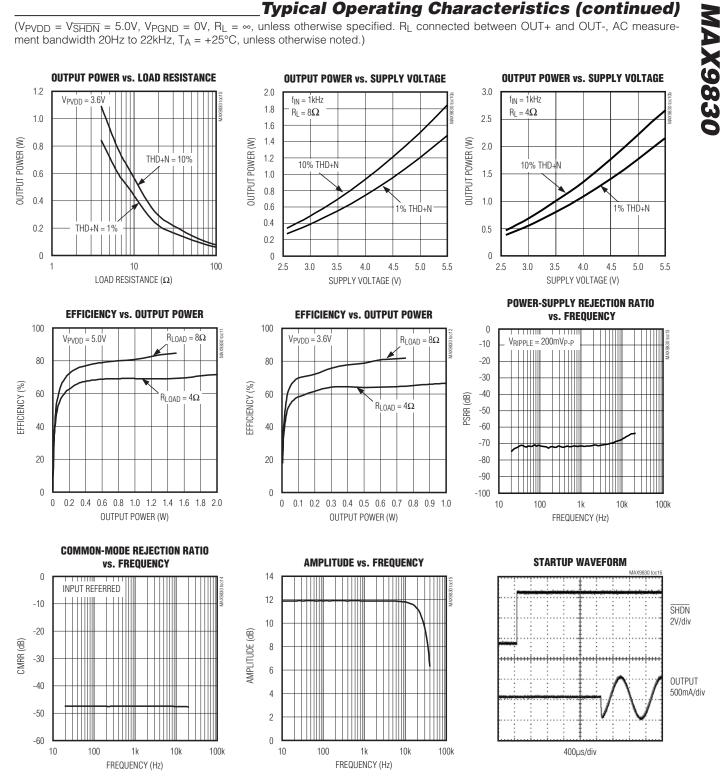
TOTAL HARMONIC DISTORTION

100k

1.4

100

4



Typical Operating Characteristics (continued) (V_{PVDD} = V_{SHDN} = 5.0V, V_{PGND} = 0V, R_L = ∞, unless otherwise specified. R_L connected between OUT+ and OUT-, AC measurement bandwidth 20Hz to 22kHz, $T_A = +25^{\circ}C$, unless otherwise noted.) SHUTDOWN WAVEFORM WIDEBAND OUTPUT SPECTRUM 0 RBW = 100Hz-20 SHDN OUTPUT AMPLITUDE (dBV) 2V/div -40 -60 OUTPUT -80 500mA/div -100 -120 400µs/div 10 100 1000 1 FREQUENCY (MHz) SUPPLY CURRENT **OUTPUT FREQUENCY SPECTRUM** vs. SUPPLY VOLTAGE 0 2.0 -20 1.6 OUTPUT MAGNITUDE (dBV) SUPPLY CURRENT (mA) -40 1.2 -60 -80 0.8 -100 0.4 -120 -140 0 2 0 4 6 8 10 12 14 16 18 20 2.5 3.0 3.5 4.0 4.5 5.0 5.5 FREQUENCY (kHz) SUPPLY VOLTAGE (V) SHUTDOWN CURRENT vs. SUPPLY VOLTAGE **RF IMMUNITY vs. FREQUENCY** 0.6 0 -10 0.5 SHUTDOWN CURRENT (µA) -20 RF IMMUNITY (dBV) 0.4 -30 0.3 -40 -50 0.2 -60 0.1 -70 0 -80 2.5 3.0 3.5 4.0 4.5 5.0 5.5 500 1000 1500 2000 3000 0 2500 SUPPLY VOLTAGE (V) FREQUENCY (MHz)

M/IXI/M

6

MAX9830

Pin Description

PIN	NAME	FUNCTION
1	IN+	Noninverting Audio Input
2	IN-	Inverting Audio Input
3	SHDN	Active-Low Shutdown Input. Drive SHDN low to place the device in shutdown mode.
4	N.C.	No Connection. Leave unconnected.
5	PGND	Ground
6	OUT-	Negative Speaker Output
7	OUT+	Positive Speaker Output
8	PVDD	Power Supply. Bypass PVDD to PGND with a 0.1µF capacitor.
	EP	Exposed Pad. Connect exposed pad to a solid ground plane.

Detailed Description

The MAX9830 features industry-leading quiescent current, low-power shutdown mode, comprehensive clickand-pop suppression, and excellent RF immunity.

The MAX9830 offers Class AB audio performance with Class D efficiency in a minimal board-space solution. The Class D amplifier features spread-spectrum modulation combined with edge rate and overshoot control circuitry that offers significant improvements to switch-mode amplifier radiated emissions.

The MAX9830 includes thermal overload and short-circuit protection.

Class D Speaker Amplifier

The MAX9830 filterless Class D amplifier offers much higher efficiency than Class AB amplifiers. The high efficiency of a Class D amplifier is due to the switching operation of the output stage transistors. Any power loss associated with the Class D output stage is mostly due to the I²R loss of the MOSFET on-resistance and quiescent current overhead.

Ultra-Low EMI Filterless Output Stage

Traditional Class D amplifiers require the use of external LC filters, or shielding, to meet EN55022B electromagnetic-interference (EMI) regulation standards. Maxim's patented active emissions limiting edge-rate control circuitry and spread-spectrum modulation reduces EMI emissions, while maintaining up to 85% efficiency.

Maxim's patented spread-spectrum modulation mode flattens wideband spectral components, while proprietary techniques ensure that the cycle-to-cycle variation of the switching period does not degrade audio reproduction or efficiency. The MAX9830's spreadspectrum modulator randomly varies the switching



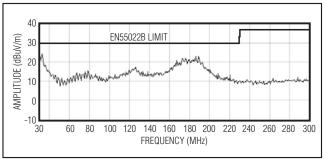


Figure 1. EMI with 24in of Speaker Cable

frequency by ± 10 kHz around the center frequency (600kHz). Above 10MHz, the wideband spectrum looks like noise for EMI purposes (Figure 1).

Speaker Current Limit

If the output current of the speaker amplifier exceeds the current limit (1.8A typ), the MAX9830 disables the outputs for approximately 400µs. At the end of 400µs, the outputs are re-enabled. If the fault condition still exists, the MAX9830 continues to disable and re-enable the outputs until the fault condition is removed.

Shutdown

The MAX9830 features a low-power shutdown mode, drawing 0.5 μ A of supply current. Drive SHDN low to put the MAX9830 into shutdown.

Click-and-Pop Suppression

The MAX9830 speaker amplifier features Maxim's comprehensive click-and-pop suppression. During startup, the click-and-pop suppression circuitry reduces any audible transient sources internal to the device. When entering shutdown, the differential speaker outputs ramp down to PGND quickly and simultaneously.

Applications Information

Filterless Class D Operation

Traditional Class D amplifiers require an output filter. The filter adds cost, size, and decreases efficiency and THD+N performance. The MAX9830's filterless modulation scheme does not require an output filter.

Because the switching frequency of the MAX9830 is well beyond the bandwidth of most speakers, voice coil movement at the switching frequency is very small. Use a speaker with a series inductance > 10μ H. Typical 8Ω speakers exhibit series inductances in the 20μ H to 100μ H range.

Component Selection

Optional Ferrite Bead Filter

Although not normally needed, in applications where speaker leads exceed 24in at VPVDD = 3V, use a filter constructed from an inexpensive ferrite bead and a small-value capacitor to ground (Figure 2) to provide additional EMI suppression. Use a ferrite bead with low DC resistance, high frequency (\geq 1MHz) impedance of 100 Ω to 600 Ω , and rated for at least 1A. The capacitor value varies based on the ferrite bead chosen and the actual speaker lead length. Select the capacitor value based on EMI performance.

Speaker Amplifier Power Supply Input (PVDD)

PVDD powers the speaker amplifier. PVDD ranges from 2.6V to 5.5V. Bypass PVDD with a 0.1μ F capacitor to PGND. Apply additional bulk capacitance at the device if long input traces between PVDD and the power source are used. Ensure a rate of voltage rise at PVDD is limited to $1V/\mu$ s.

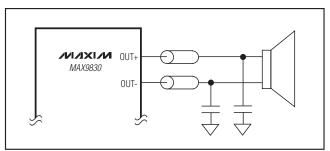


Figure 2. Optional Ferrite Bead Filter

Input Filtering

The input-coupling capacitor (C_{IN}), in conjunction with the amplifier's internal input resistance (R_{IN}), forms a highpass filter that removes the DC bias from the incoming signal. These capacitors allow the amplifier to bias the signal to an optimum DC level. Select 0.47μ F capacitors for optimum click-and-pop performance and 17Hz f-3dB.

If a different $f_{\text{-}3dB}$ is required, $C_{\text{IN},}$ assuming zero-source-impedance, is:

$$C_{IN} = \frac{8}{f_{-3dB}} [\mu F]$$

Use capacitors with adequately low voltage-coefficient for best low-frequency THD performance.

Layout and Grounding

Proper layout and grounding are essential for optimum performance. Good grounding improves audio performance and prevents switching noise from coupling into the audio signal.

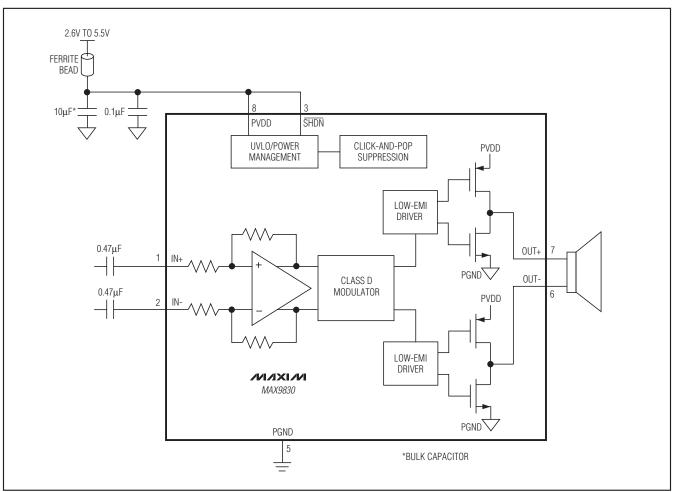
Use wide, low-resistance output traces. As load impedance decreases, the current drawn from the device outputs increase. At higher current, the resistance of the output traces decrease the power delivered to the load. For example, if 2W is delivered from the speaker output to a 4 Ω load through a 100m Ω trace, 49mW is consumed in the trace. If power is delivered through a 10m Ω trace, only 5mW is consumed in the trace. Wide output, supply and ground traces also improve the power dissipation of the device.

The MAX9830 is inherently designed for excellent RF immunity. For best performance, add ground fills around all signal traces on top and bottom PCB planes.

The MAX9830 TDFN package features an exposed thermal pad on its underside. This pad lowers the package's thermal resistance by providing a heat conduction path from the die to the PCB. Connect the exposed thermal pad to the ground plane by using a large pad and multiple vias.

Chip Information

PROCESS: CMOS



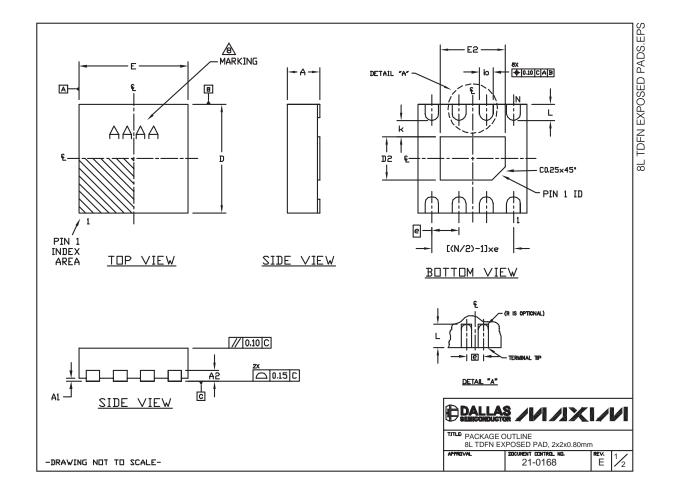
_Functional Diagram

MAX9830

Package Information

For the latest package outline information and land patterns, go to <u>www.maxim-ic.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	DOCUMENT NO.
8 TDFN-EP	T822+2	<u>21-0168</u>





Package Information (continued)

For the latest package outline information and land patterns, go to <u>www.maxim-ic.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

COMMO	ON DIMENSI	ONS						
SYMBOL	MIN.	MAX.						
A	0.70	0.80						
D	1.90	2.10						
E	1.90	2.10						
A1	0.00	0.05						
L	0.20	0.40						
k	0.25 M							
A2	0.20 F	REF.						
PACKAGE	VARIATION	s						7
PKG. CODI		D2	E2	е	b	r	[(N/2)-1] x e	-
T822-1	8	0.70±0.10	1.30±0.10	0.50 TYP.	0.25±0.05	0.125	1.50 REF	-
T822-2	8	0.80±0.10	1.20±0.10	0.50 TYP.	0.25±0.05	0.125	1.50 REF	7
L							1	
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MAX9830

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	8/09	Initial release	_
1	4/10	Removed PSRR spec from the <i>Features</i> section, updated EC table specs, and added new TOCs	1, 2, 5

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Revision History